

# Data-Driven Control via Semidefinite Programming in Nonlinear Systems

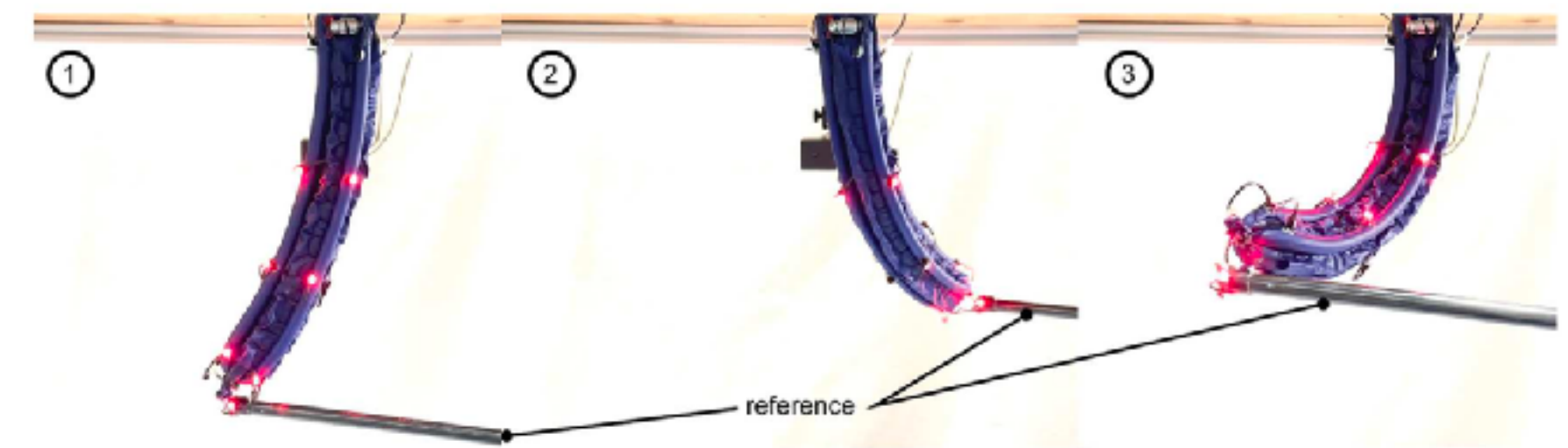
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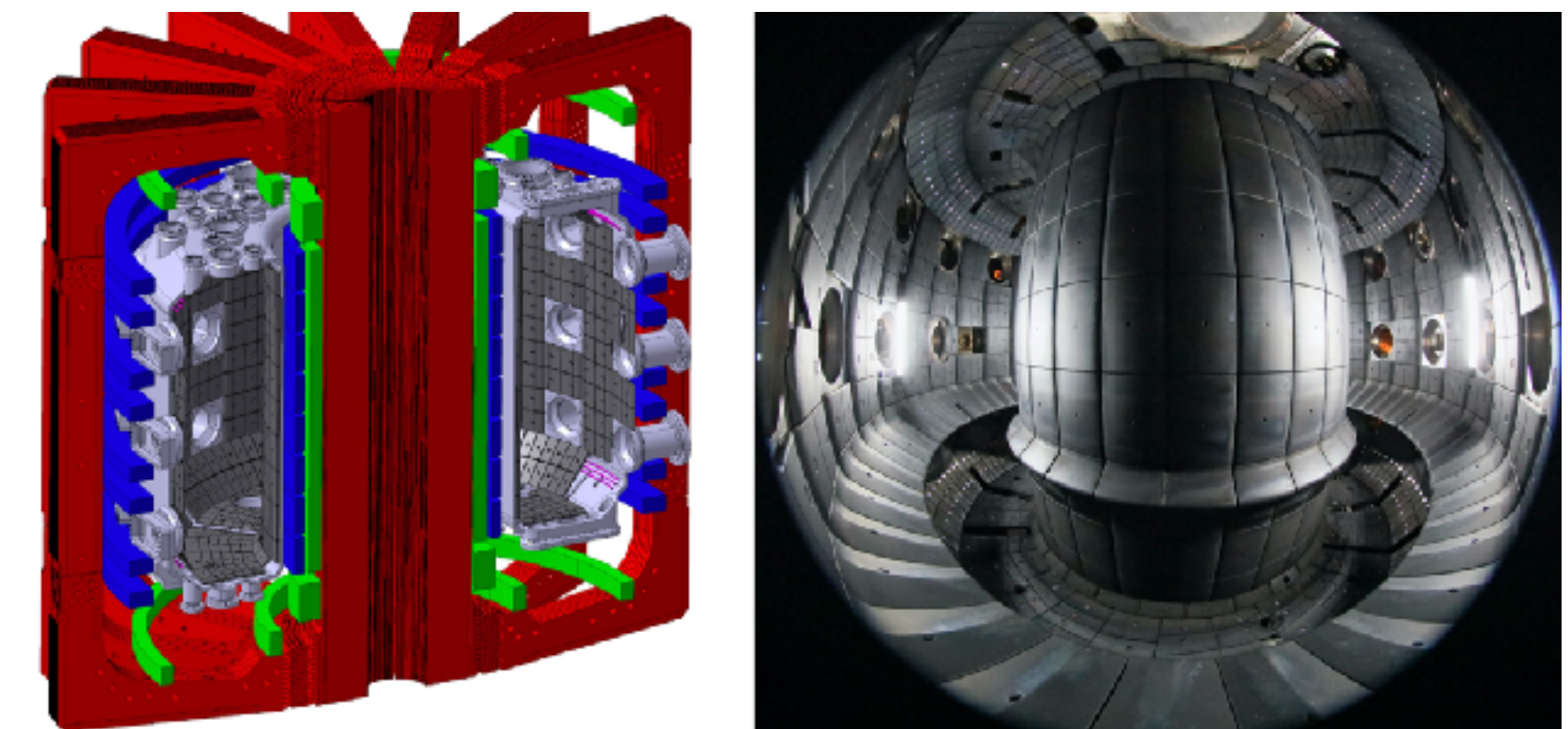
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# Data-Driven Frameworks for Control

- Data-driven control methods are useful for tackling complex engineering systems
  - Working with data vs. complex models
- Soft robotics, tokamaks for fusion energy, power grids have already seen success<sup>1,2,3</sup>
- Developing these methods rigorously is important for both intelligent and practical application
  - Semidefinite programming can leverage assumptions about the system for stability guarantees



Pneumatically actuated soft arm<sup>1</sup>



Tokamak à Configuration Variable<sup>2</sup>

[1] Haggerty, D. A., Banks, M. J., Kamenar, E., Cao, A. B., Curtis, P. C., Mezić, I., & Hawkes, E. W. (2023). Control of soft robots with inertial dynamics. *Science Robotics*, 8(81)

[2] Degraeve, J., Felici, F., Buchli, J., Neunert, M., Tracey, B., Carpanese, F., Ewalds, T., Hafner, R., Abdolmaleki, A., De Las Casas, D., Donner, C., Fritz, L., Galperti, C., Huber, A., Keeling, J., Tsimpoukelli, M., Kay, J., Merle, A., Moret, J.-M., ... Riedmiller, M. (2022). Magnetic control of tokamak plasmas through deep reinforcement learning. *Nature*, 602(7897), 414–419

[3] Yuan, Z., Zhao, C., & Cortés, J. (2024). Reinforcement learning for distributed transient frequency control with stability and safety guarantees. *Systems & Control Letters*, 185



# Sum of Squares Polynomials

**Definition** (Sum of squares polynomials): A polynomial  $p(x)$  of degree  $2d$  ( $d \in \mathbb{Z}_{>0}$ ) is sum of squares (SOS) if there exist polynomials  $g_1(x), \dots, g_k(x)$  s.t.

$$p(x) = \sum_{i=1}^k g_i(x)^2$$

$$p(x) \text{ SOS} \implies p(x) \text{ nonnegative } \forall x \neq 0$$

**Theorem** (Representation via PSD matrices) [4]: A polynomial  $p(x)$  of degree  $2d$  is SOS if and only if there exists a matrix  $Q \succ 0$  of appropriate dimension such that with  $m(x)$  vector consisting of all monomials in  $x$  of degree at most  $d$ ,

$$p(x) = m(x)^\top Q m(x)$$

Ex:  $2x_1^2 + 2x_1x_2 + 2x_2^2 = [x_1 \ x_2] \begin{bmatrix} 2 & 1 \\ 1 & 2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$

[4] Nesterov, Y. (2000). Squared Functional Systems and Optimization Problems. In H. Frenk, K. Roos, T. Terlaky, & S. Zhang (Eds.), *High Performance Optimization* (Vol. 33, pp. 405–440)

# Semidefinite Programs for Data-Driven Control of Linear Systems

Given discrete-time linear system we would like to stabilize to origin:

Suppose  $A, B$  unknown, but snapshot/trajectory data can be generated:

$$x(t+1) = Ax(t) + Bu(t)$$

$$\mathcal{U} = [u(0) \quad u(1) \quad \cdots \quad u(T-1)]$$

$$\mathcal{X}_- = [x(0) \quad x(1) \quad \cdots \quad x(T-1)]$$

$$\mathcal{X}_+ = [x(1) \quad x(2) \quad \cdots \quad x(T)]$$

**Theorem** [5]: The data  $(\mathcal{U}, \mathcal{X}_-, \mathcal{X}_+)$  are informative for stabilization by state feedback if and only if there exists a matrix  $\Theta \in \mathbb{R}^{T \times n}$  satisfying:

$$\mathcal{X}_- \Theta = (\mathcal{X}_- \Theta)^\top \text{ and } \begin{bmatrix} \mathcal{X}_- \Theta & \mathcal{X}_+ \Theta \\ \Theta^\top \mathcal{X}_+^\top & \mathcal{X}_- \Theta \end{bmatrix} \succ 0$$

Moreover,  $K = \mathcal{U} \Theta (\mathcal{X}_- \Theta)^{-1}$  will stabilize the system.

[5] van Waarde, H. J., Eising, J., Trentelman, H. L., & Camlibel, M. K. (2020). Data Informativity: A New Perspective on Data-Driven Analysis and Control. *IEEE Transactions on Automatic Control*, 65(11), 4753–4768. [IEEE Transactions on Automatic Control](#).

# How Assumptions on Underlying Dynamics Grant Stability Guarantees

Model-based control design:

System  
matrices  
 $(A, B)$



Choose  $K$  such  
that  $(A + BK)$   
Hurwitz/Schur



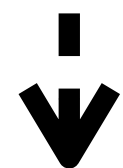
$$P(A + BK) + (A + BK)^{\top} P < 0$$

or

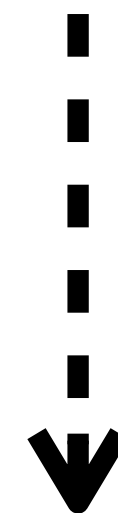
$$(A + BK)^{\top} P(A + BK) - P < 0$$



Data-based control design:



Trajectory  
data  
 $(\mathcal{U}, \mathcal{X}_-, \mathcal{X}_+)$



Parameterize  $(A + BK)$   
with data and solve  
requirements with data

# Data Efficiency of Direct Control Design

- The previous result shows how data can be used to directly design control, **bypassing model identification**
  - The method searches for a gain matrix  $K$  that stabilizes the collection of systems consistent with the data
  - **Requires less data** than identification
- Willems's fundamental lemma and persistency of excitation<sup>6,7</sup> give a precise answer for how the data should be collected

[6] Willems, J. C., Rapisarda, P., Markovsky, I., & De Moor, B. L. M. (2005). A note on persistency of excitation. *Systems & Control Letters*, 54(4), 325–329.

[7] De Persis, C., & Tesi, P. (2020). Formulas for Data-Driven Control: Stabilization, Optimality, and Robustness. *IEEE Transactions on Automatic Control*, 65(3), 909–924. *IEEE Transactions on Automatic Control*.

# Problem Setup for Nonlinear Case

Given nonlinear system where dynamics unknown, but we can collect data:

$$\dot{x}(t) = f(x) + Bu, \quad x \in \mathbb{R}^n, \quad u \in \mathbb{R}^m \quad (1)$$

$$\mathcal{U} = [u(0) \quad u(1) \quad \cdots \quad u(T-1)]$$

$$\mathcal{X}_- = [x(0) \quad x(1) \quad \cdots \quad x(T-1)]$$

$$\mathcal{X}_d = [\dot{x}(0) \quad \dot{x}(1) \quad \cdots \quad \dot{x}(T-1)]$$

Assume we have a dictionary of functions  $Z(x) : \mathbb{R}^n \rightarrow \mathbb{R}^S$  that can be used to represent  $f(x)$ :

$$\dot{x}(t) = AZ(x) + Bu \quad (\text{where } A \in \mathbb{R}^{n \times S} \text{ unknown})$$

Can compute “lifted” data:

$$\mathcal{Z}_- = [Z(x(0)) \quad Z(x(1)) \quad \cdots \quad Z(x(T-1))]$$

Data is consistent with the equation:

$$\mathcal{X}_d = A\mathcal{Z}_- + B\mathcal{U}$$

[8] Guo, M., De Persis, C., & Tesi, P. (2020). Learning control for polynomial systems using sum of squares relaxations. *2020 59th IEEE Conference on Decision and Control (CDC)*, 2436–2441.



# Design of Control and Lyapunov Function

Assume control has form  $u = KZ(x)$ .  
Closed loop system is:

$$\dot{x} = (A + BK)Z(x)$$

Consider candidate Lyapunov  
function:

$$V(x) = Z(x)^\top P Z(x) \text{ with } P \succ 0$$

$$\begin{aligned} \dot{V}(x) &= \left( \frac{\partial Z(x)}{\partial x} (A + BK) Z(x) \right)^\top P Z(x) + Z(x)^\top P \left( \frac{\partial Z(x)}{\partial x} (A + BK) Z(x) \right) \\ &= Z(x)^\top \left[ \left( \frac{\partial Z(x)}{\partial x} (A + BK) \right)^\top P + P \frac{\partial Z(x)}{\partial x} (A + BK) \right] Z(x) \end{aligned}$$

[8] Guo, M., De Persis, C., & Tesi, P. (2020). Learning control for polynomial systems using sum of squares relaxations. *2020 59th IEEE Conference on Decision and Control (CDC)*, 2436–2441.



# Rewriting Expressions in Terms of Data

$$\dot{V}(x) = Z(x)^\top \left[ \left( \frac{\partial Z(x)}{\partial x} (A + BK) \right)^\top P + P \frac{\partial Z(x)}{\partial x} (A + BK) \right] Z(x)$$

To write in terms of data, assume that the data  $\mathcal{Z}_-$  is full row rank:

$$\exists G \in \mathbb{R}^{T \times S} \text{ s.t. } I_S = \mathcal{Z}_- G$$

Further, choose  $K = \mathcal{U}G$

Then, the data can parameterize the closed loop dynamics of the system under this feedback:

$$A + BK = [B \quad A] \begin{bmatrix} K \\ I_n \end{bmatrix} = [B \quad A] \begin{bmatrix} \mathcal{U} \\ \mathcal{Z}_- \end{bmatrix} G = \mathcal{X}_d G$$

[8] Guo, M., De Persis, C., & Tesi, P. (2020). Learning control for polynomial systems using sum of squares relaxations. *2020 59th IEEE Conference on Decision and Control (CDC)*, 2436–2441.

# Tightened Conditions for Negative Definiteness

$$\dot{V}(x) = Z(x)^\top P \left[ P^{-1} G^\top \mathcal{X}_d^\top \left( \frac{\partial Z(x)}{\partial x} \right)^\top + \frac{\partial Z(x)}{\partial x} \mathcal{X}_d G P^{-1} \right] P Z(x)$$

Then to write conditions for design of stabilizing control, we wish to impose negative definiteness of  $\dot{V}(x)$  and search for valid matrices  $G, P$  satisfying all previous conditions

However, solving this computationally is not trivial due to the bilinearity in  $G, P$

A common sufficient requirement is to impose negative definiteness of:

$$P^{-1} G^\top \mathcal{X}_d^\top \left( \frac{\partial Z(x)}{\partial x} \right)^\top + \frac{\partial Z(x)}{\partial x} \mathcal{X}_d G P^{-1}$$

[8] Guo, M., De Persis, C., & Tesi, P. (2020). Learning control for polynomial systems using sum of squares relaxations. *2020 59th IEEE Conference on Decision and Control (CDC)*, 2436–2441.

# Sufficient Conditions for Nonlinear Stabilization

$$-P^{-1}G^{\top}\mathcal{X}_d^{\top}\left(\frac{\partial Z(x)}{\partial x}\right)^{\top}-\frac{\partial Z(x)}{\partial x}\mathcal{X}_dGP^{-1}\succ 0, \quad \forall x \in \mathbb{R}^n$$

Letting  $Y = GP^{-1}$ , we can collect the results in the following theorem:

**Theorem [8]:** For the nonlinear system (1), assuming  $Z(x) = 0 \iff x = 0$  and full row rank data matrix  $\mathcal{Z}_-$ , if there exist matrices  $Y \in \mathbb{R}^{T \times S}$ ,  $P \in \mathbb{R}^{S \times S}$ ,  $P \succ 0$  and SOS poly  $\epsilon(x)$  such that:

1.  $\mathcal{Z}_-Y = P^{-1}$  (because  $\mathcal{Z}_-Y = \mathcal{Z}_-GP^{-1} = I_S P^{-1} = P^{-1}$ )
2.  $-Y^{\top}\mathcal{X}_d^{\top}\left(\frac{\partial Z(x)}{\partial x}\right)^{\top}-\frac{\partial Z(x)}{\partial x}\mathcal{X}_dY - \epsilon(x)I_S$  is an SOS matrix polynomial

Then  $u(x) = \mathcal{U}YP^{-1}Z(x) = \mathcal{U}GZ(x)$  stabilizes the polynomial system globally.

[8] Guo, M., De Persis, C., & Tesi, P. (2020). Learning control for polynomial systems using sum of squares relaxations. *2020 59th IEEE Conference on Decision and Control (CDC)*, 2436–2441.

# Key Assumptions That Obstruct Necessity

- The previous result is a sufficient result relies on:
  - Assuming there will exist Lyapunov functions of form  $Z(x)^\top P Z(x)$
  - Tightening negative definiteness of  $Z(x)^\top P \begin{bmatrix} H(x) \end{bmatrix} P Z(x)$  to  $H(x) < 0 \quad \forall x$
- Essentially, **if the method fails, nothing is known about whether or not the system was stabilizable** via different approaches
- Necessary conditions in the form of feasibility guarantees are critical
  - Classes of nonlinear systems where any challenges resolved?



# Systems that Permit Linear Embeddings

Consider nonlinear systems that permit linear embeddings, where  $\dot{x} = f(x) = AZ(x)$  and  $\exists \phi(x) : \mathbb{R}^n \rightarrow \mathbb{R}^W$  such that  $\dot{\phi}(x) = \tilde{A}\phi(x)$  (where both  $\phi(x)$ ,  $\tilde{A}$  unknown)

Assuming  $\phi(x) = TZ(x)$ ,  $T \in \mathbb{R}^{W \times S}$   
(our dictionary contains the components  
of  $\phi(x)$ ):

$$\begin{aligned}\dot{\phi}(x) &= T\dot{Z}(x) = T \frac{\partial Z(x)}{\partial x} AZ(x) \\ &= \tilde{A}\phi(x) = \tilde{A}TZ(x)\end{aligned}$$

$$\implies Z(x) \in \mathcal{N} \left( \tilde{A}T - T \frac{\partial Z(x)}{\partial x} A \right)$$

$$\exists P \text{ s.t. } \dot{V}(x) = Z(x)^\top \left[ A^\top \left( \frac{\partial Z(x)}{\partial x} \right)^\top P + P \frac{\partial Z(x)}{\partial x} A \right] Z(x) = Z(x)^\top [-T^\top T] Z(x)$$

However, issues persist...

# Exponentially Stable Polynomial Systems Offer Useful Converse Lyapunov Theorems

- Polynomial systems that are exponentially stable on compact sets have Lyapunov functions that are SOS polynomials and have bounded degree<sup>9</sup>
  - This addresses some aspects of the dictionary question and Lyapunov functions of the form  $Z(x)^{\top} P Z(x)$
  - However, no guarantees given on whether or not  $-\dot{V}(x)$  is SOS<sup>10</sup> and once again,  $Z(x)^{\top} [H(x)] Z(x) > 0 \not\Rightarrow H(x) \succ 0$

[9] Peet, M. M., & Papachristodoulou, A. (2012). A Converse Sum of Squares Lyapunov Result With a Degree Bound. *IEEE Transactions on Automatic Control*, 57(9), 2281–2293.

[10] Ahmadi, A. A., & Parrilo, P. A. (2011). Converse results on existence of sum of squares Lyapunov functions. *IEEE Conference on Decision and Control and European Control Conference*, 6516–6521.

# Concluding Remarks

- Data-driven approaches for control have shown promise in designing control for systems that can be difficult to model
- However, developing these methods with the same rigor as in model-based control is critical
- While semidefinite programming has yielded insights into how data-driven control can be characterized for linear systems, **extensions to nonlinear systems remain very challenging**





# Numerical Example

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} x_2 + x_1^3 + u \\ -x_1 \end{bmatrix} \quad Z(x) = \begin{bmatrix} x_1 & x_2 & x_1^2 & x_2^2 & x_1^3 \end{bmatrix}^T$$

Collecting 9 data points with forcing  $u = -\sin(t)$ , the previous method yields:

$$u = KZ(x) = -8.5315x_1 - x_2 - x_1^3$$

$$P = 1.0e+08 * \begin{bmatrix} 0.000000002464676 & 0.000000005247414 & -0.000000000656395 & -0.000000000386742 & 0.000000000213237 \\ 0.000000005247414 & 0.462610683393534 & 0.334319457507848 & 0.657146896297293 & -0.323265298534464 \\ -0.000000000656395 & 0.334319457507848 & 0.497647391617901 & 0.307730545046117 & -0.202924820664369 \\ -0.000000000386742 & 0.657146896297293 & 0.307730545046117 & 1.086556755963401 & -0.443877017893051 \\ 0.000000000213237 & -0.323265298534464 & -0.202924820664369 & -0.443877017893051 & 0.264821810384319 \end{bmatrix}$$

